Massachusetts Institute of Technology Department of Electrical Engineering and Computer Science

6.341: DISCRETE-TIME SIGNAL PROCESSING

Fall 2005

Problem Set 11

Hilbert Transforms, Cepstral Analysis

Issued: Thursday, December 1, 2005.

 $\mathbf{Due:}\ \mathrm{Never}$

Problem Set 11: Although this problem set will not be graded (Institute regulations), in preparing the final exam we assume that you will have done Problem Set 11 and, of course, all the others. Solutions to Problem Set 11 will be handed out in the last lecture on December 8.

Schedule: Because of the final exam schedule, the last lecture will be on Thursday, December 8. There will be no lecture on Tuesday, December 13. The final exam is on Friday, December 16 from 1:30 PM - 4:30 PM.

Reading: Lecture 21 on December 1 discusses Hilbert transforms. The relevant reading in the text is sections 11.1, 11.3, and 11.4. Lecture 22 on December 6 is on cepstral analysis. Cepstral analysis is discussed in the draft chapter to be handed out in lecture. The primary sections to read are 12.1 through 12.6. You should also at least look through the application examples in 12.9 and 12.10.

Problem 11.1

OSB Problem 11.3

Note: The answers in the back of the book may not be correct in your version of the textbook.

Problem 11.2

y[n] is a real, causal sequence with DTFT $Y(e^{j\omega})$. Determine y[n] if

$$\operatorname{Im}\{Y(e^{j\omega})\} = 3\sin\omega + \sin 3\omega,$$

and

$$Y(e^{j\omega})\big|_{\omega=\pi} = 3.$$

Problem 11.3

OSB Problem 11.5

Problem 11.4

The system shown in Figure 11.4-1 is referred to as a 90° phase splitter.

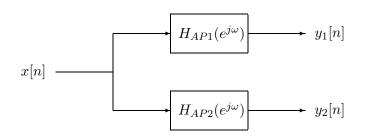


Figure 11.4-1: A 90° phase splitter.

The two all-pass systems $H_{AP1}(e^{j\omega})$ and $H_{AP2}(e^{j\omega})$ are designed so that their phases differ by $\pi/2$:

$$H_{AP1}(e^{j\omega}) = e^{j\theta(\omega)}, \quad -\pi < \omega < \pi$$
$$H_{AP2}(e^{j\omega}) = \begin{cases} e^{j(\theta(\omega) - \pi/2)}, & 0 < \omega < \pi\\ e^{j(\theta(\omega) + \pi/2)}, & -\pi < \omega < 0 \end{cases}$$

If we consider $y_1[n]$ and $y_2[n]$ to be the real and imaginary parts respectively of a complex signal w[n], i.e. $w[n] = y_1[n] + jy_2[n]$, show that

$$W(e^{j\omega}) = 0, \quad -\pi \le \omega < 0$$

and

$$|W(e^{j\omega})| = 2|X(e^{j\omega})|, \quad 0 < \omega < \pi$$

where $W(e^{j\omega})$ and $X(e^{j\omega})$ are the DTFTs of w[n] and x[n] respectively.

Note: You may find it helpful to read section 11.4 before doing this problem.

Problem 11.5

The sequence

$$h[n] = \delta[n] + \alpha \delta[n - n_0]$$

is a simplified model for the impulse response of a system that introduces an echo. Assume that $|\alpha| < 1$.

Determine the complex cepstrum $\hat{h}[n]$ for this sequence. Sketch the result.

Problem 11.6

Equation (12.34) in Section 12.4 of the Chapter 12 draft presents a relationship between a sequence x[n] and its complex cepstrum $\hat{x}[n]$.

$$x[n] = \sum_{k=-\infty}^{\infty} \left(\frac{k}{n}\right) \hat{x}[k]x[n-k], \quad n \neq 0$$

In addition, in Section 12.5.2 it is shown that if x[n] is minimum phase, then the complex cepstrum is zero for n < 0.

Using the above fact with x[n] being minimum phase, show that Eq. (12.34) implies:

- 1. $\hat{x}[n]$ can be computed recursively from x[n],
- 2. $\hat{x}[n]$ at any time n_0 depends only on x[n] for $n \leq n_0$ (i.e. Eq. (12.34) reduces to a causal computation).

Problem 11.7

(a) x[n] is a causal, real-valued sequence with Fourier Transform $X(e^{j\omega})$. It is known that

$$\operatorname{Re}\{X(e^{j\omega})\} = 1 + 3\cos\omega + \cos 3\omega.$$

Determine a choice for x[n] consistent with this information and specify whether or not your choice is unique.

(b) Let $\hat{x}[n]$ be the complex cepstrum of a real-valued sequence x[n].

Specify whether the following statements are true or false. Give a brief justification for your answer.

Statement 1: If $x_1[n] = x[-n]$ then $\hat{x}_1[n] = \hat{x}[-n]$.

Statement 2: Since x[n] is real-valued, the complex cepstrum $\hat{x}[n]$ must also be real-valued.